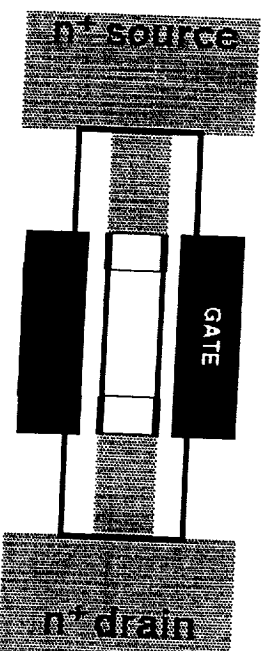


Where is scattering important in nanotransistors?

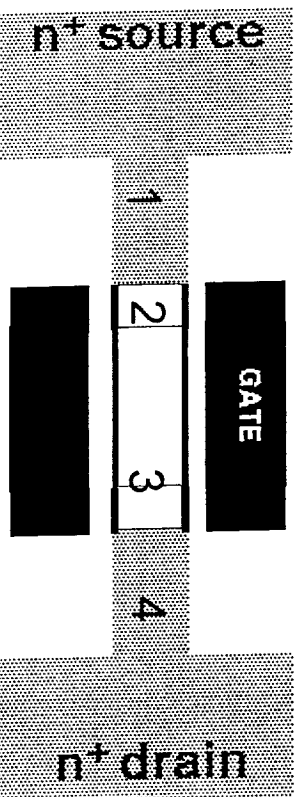
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Equations

- Developed algorithms and models for 2D (quasi-1D) simulation of nanostructures
- Equations for retarded (G^r) and less-than ($G^<$) Green's functions:
$$(E-H-\Sigma^r) G^r(r,r',E) = \delta(r-r')$$
$$(E-H-\Sigma^r) G^<(r,r',E) = \Sigma^< G^a(r,r',E)$$
$$(E-H-\Sigma^r) G^>(r,r',E) = \Sigma^> G^a(r,r',E)$$
- Σ^r represents self-energy due to open boundaries and electron-phonon scattering (self-consistent Born approx.)
- Poisson's equation

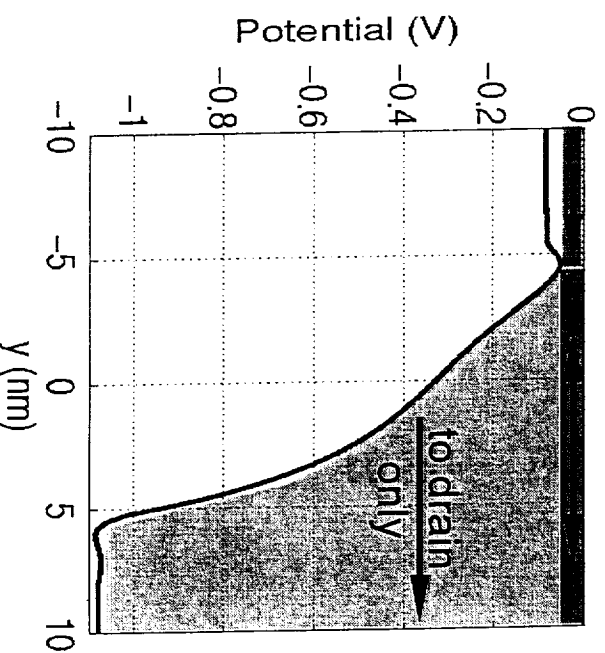


Where is scattering most detrimental to on-current? Region 1, 2, 3 & 4 below?

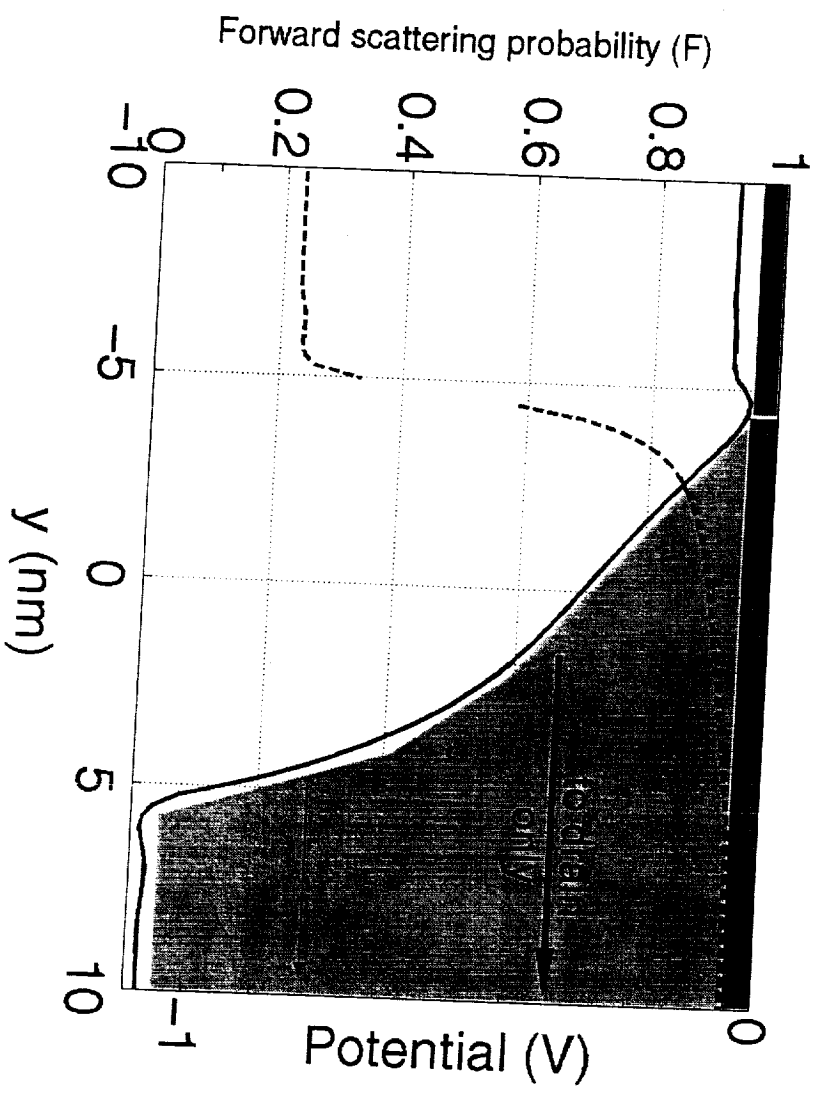


Competing Factors

1. Energy relaxation in the drain-end
2. Poisson's equation: Source injection barrier height.

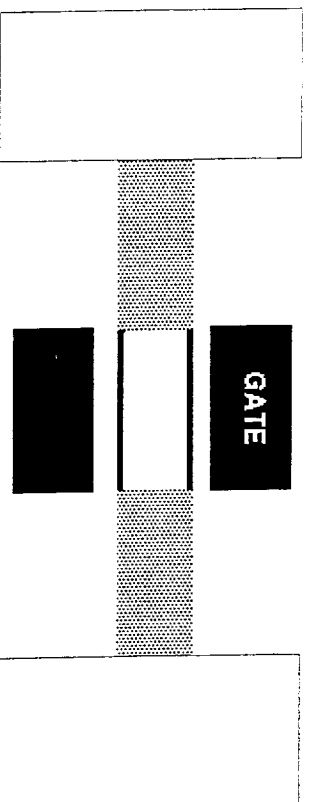


- Simple models have assumed:
 - Scattering is important only in the source-end.
 - The extension regions can be modeled as series resistances

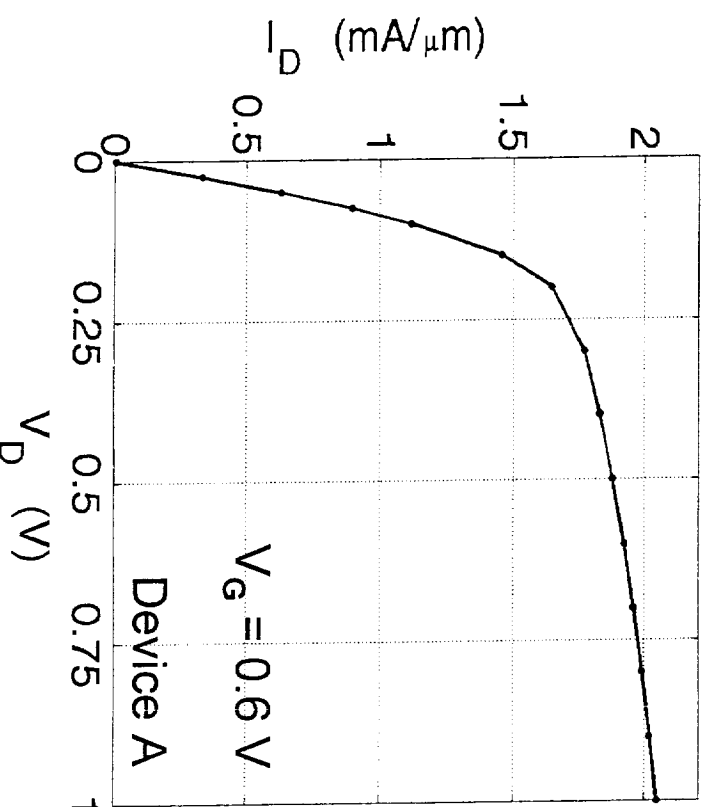


- No charge self-consistency.
- The reflected carriers contribute to channel charge. It's contribution to current is non-linear!

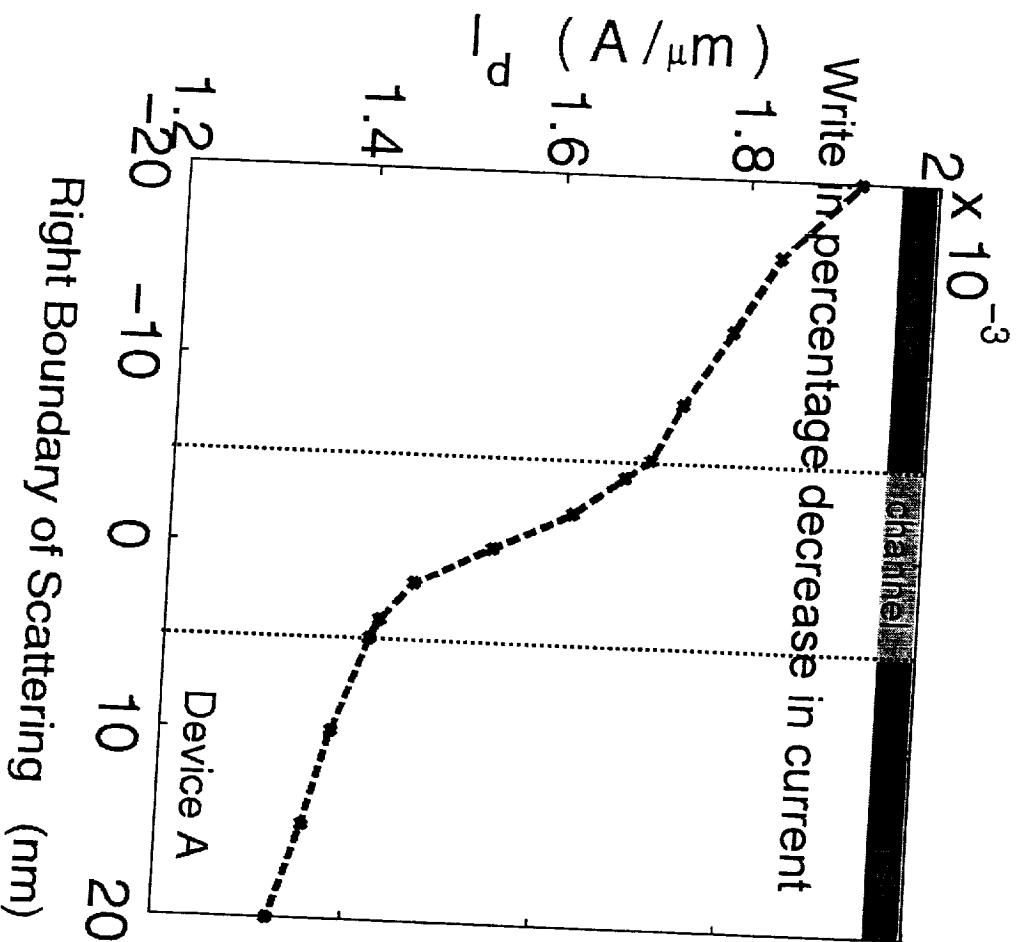
Device A: $L_{ch} = 10 \text{ nm}$



$L_{ch} = 10 \text{ nm}$ or 25 nm
 $T_{ch} = 1.5 \text{ nm}$
 $T_{ox} = 1.5 \text{ nm}$
Gate dielectric = 3.9 or 20
Doping = $1E+20 \text{ cm}^{-3}$
Channel is intrinsic

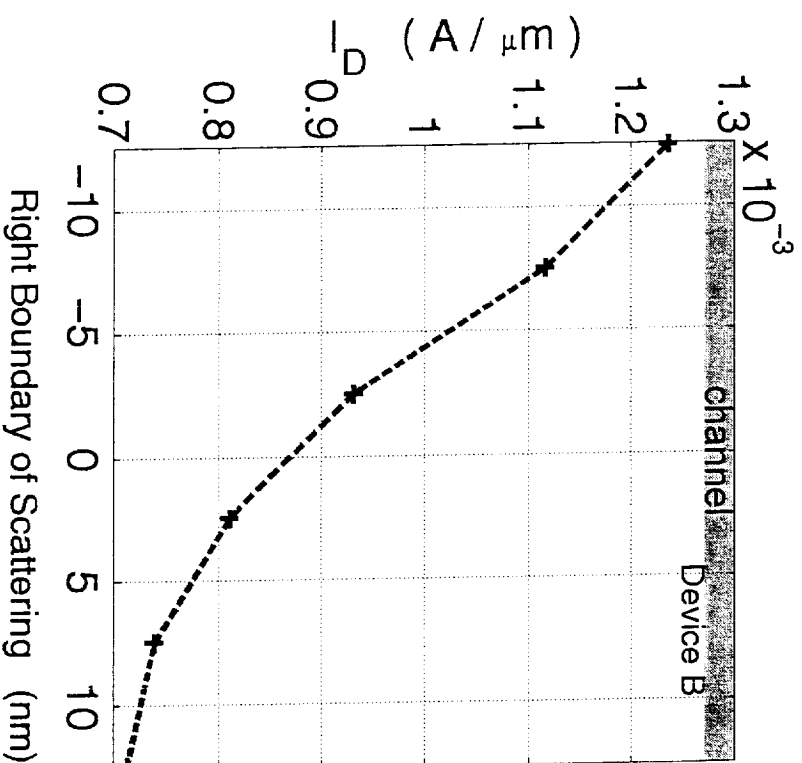


Device A: $L_{ch} = 10 \text{ nm}$



$L_{ch} = 10 \text{ nm}$
 $L_{scatt} = 10 \text{ nm}$
 Ballisticity = 68%

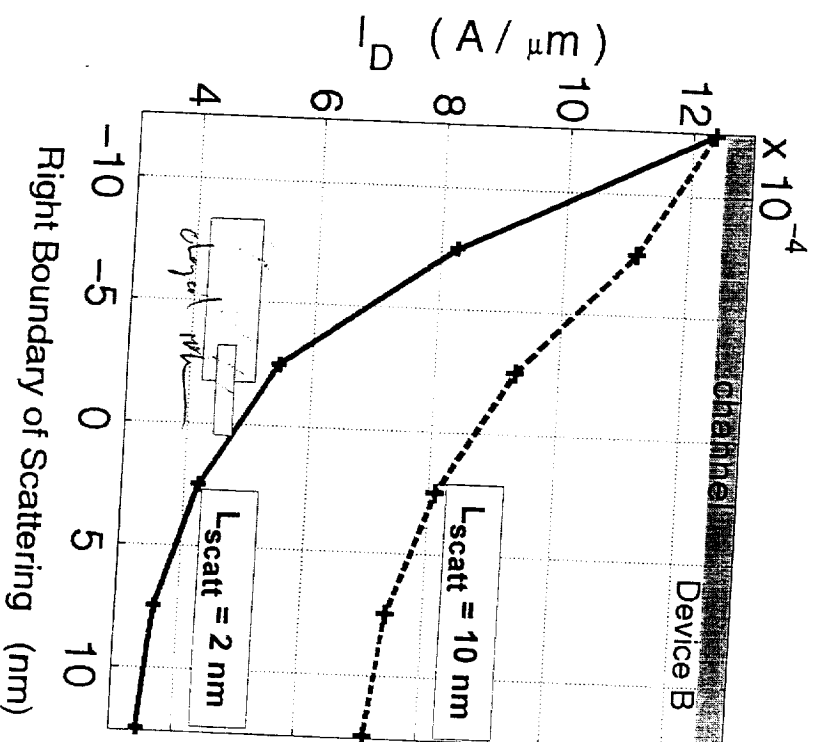
Device B: $L_{ch} = 25 \text{ nm}$



$L_{ch} = 25 \text{ nm}$
 $L_{scatt} = 10 \text{ nm}$
Ballisticity = 57%
Left Half = 69%
Right Half = 31%

- Scattering in the right half (of channel) is comparable to scattering in the left half, in causing on-current reduction.
- What happens when the scattering length is smaller?

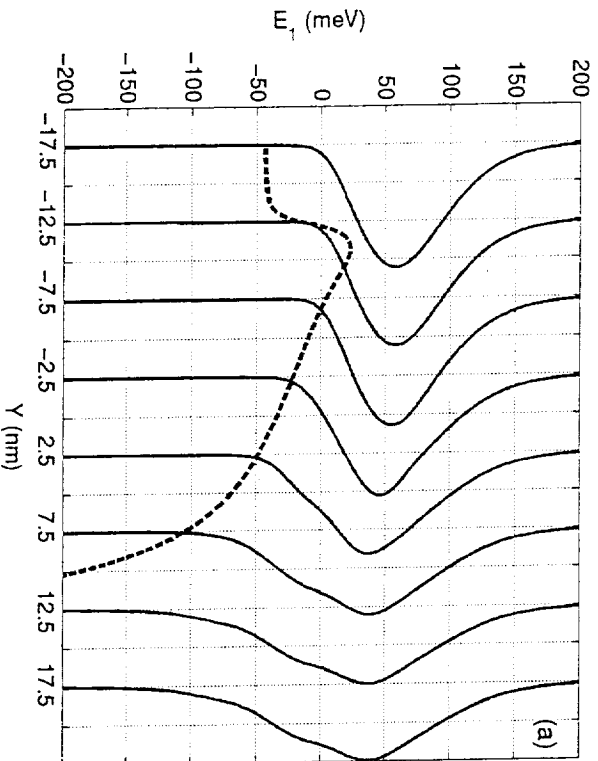
What happens when $L_{ch} \gg L_{scatt}$?



$L_{ch} = 25$ nm

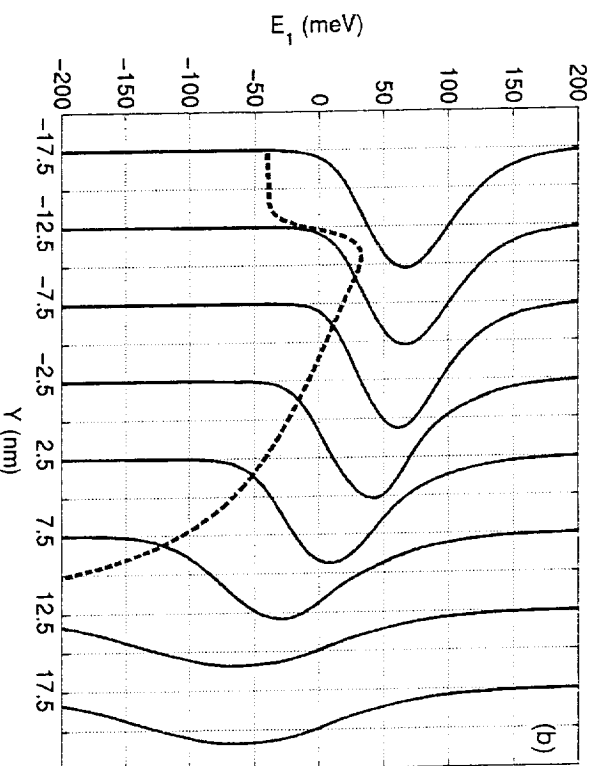
- As L_{ch} becomes much larger than L_{scatt} , scattering in the source-end becomes much more important than the drain-end

$L_{\text{scatt}} = 10 \text{ nm}$



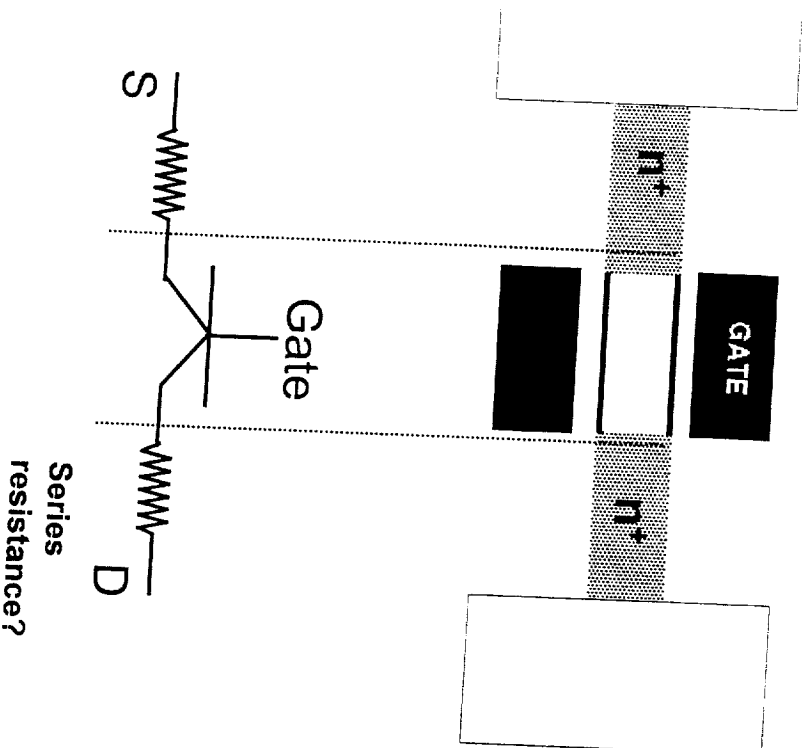
- Hot carriers at drain end
- Reflection at drain-end has a large influence

$L_{\text{scatt}} = 2 \text{ nm}$

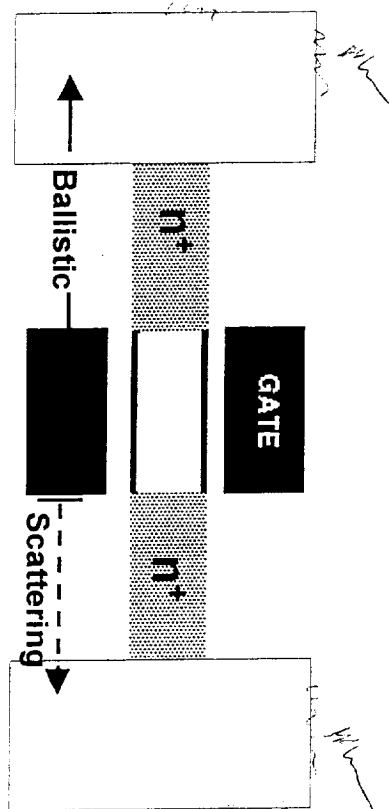


- Thermalized at drain end
- Reflection at drain-end has only a small influence

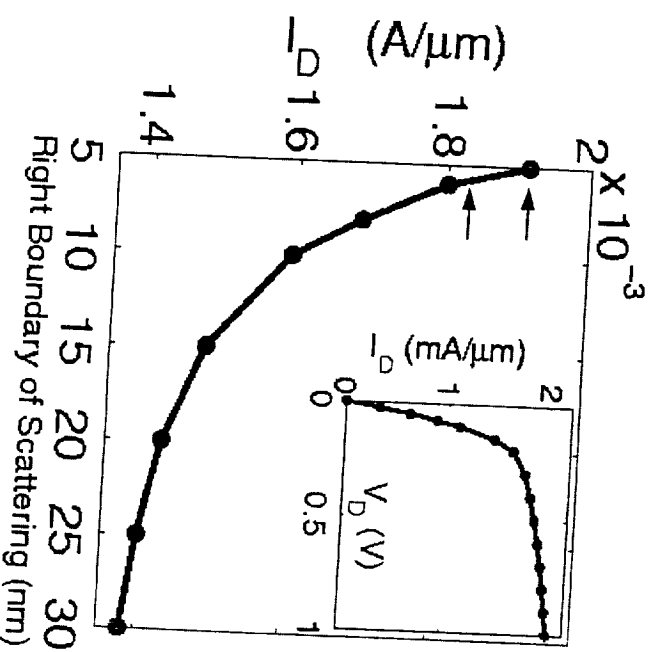
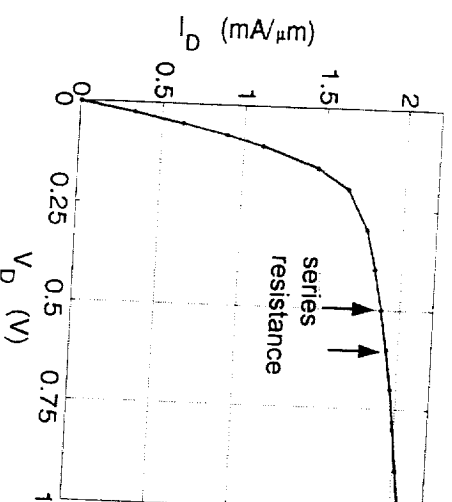
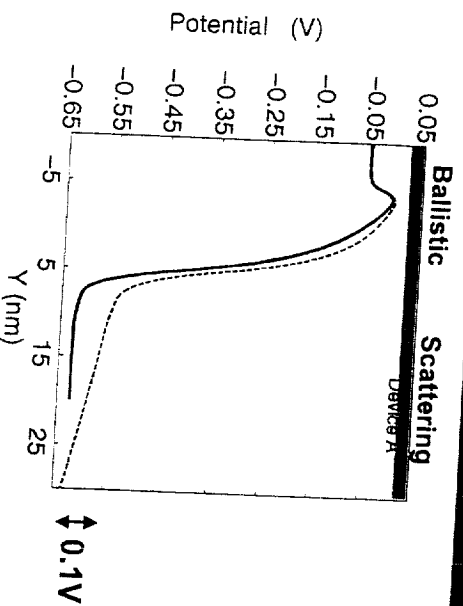
Series Resistance?

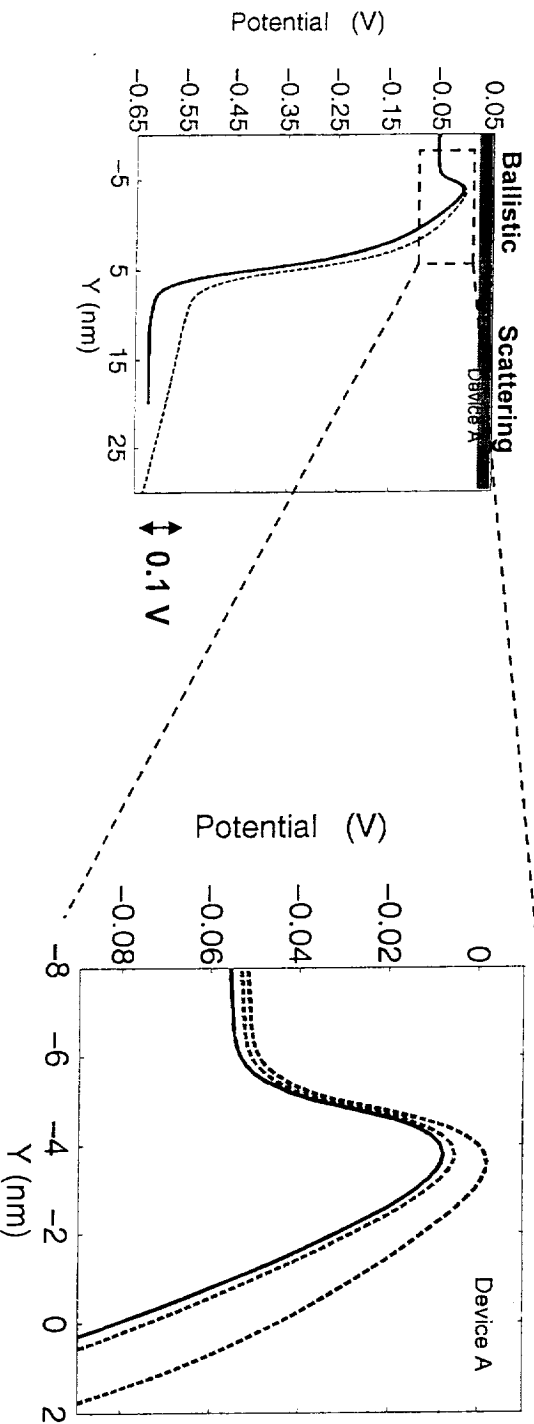


- Conventional methods treat the extension regions as a series resistance that can be added to an equivalent circuit.
- Voltage drops: $I_D R_s$ and $I_D R_D$
- $I_D [V_D, V_G] \rightarrow I_D [V_D - I_D (R_s + R_D), V_G - I_D R_s]$
- Can the highly doped extension regions be modeled as series resistances?



Potential, series resistance, current





- Source injection barrier increases due to scattering in the drain
- $V_D = 0.6 \text{ V} - 0.1 \text{ V} = 0.5 \text{ V}$ (red line)
- If the channel is ballistic, scattering in the drain extension causes a super-linear decrease in current because of electrostatics.

